

An Assessment of PR and HC controlled Bootstrap Converter Fed SVM Inverter based Induction motor drive

Lavanya R¹, Jayalakshmi V²
{maillavanya83@gmail.com, jayas_1979@yahoo.co.in}

Research Scholar, Bharath Institute of Higher Education and Research, Chennai¹
Associate Professor, Bharath Institute of Higher Education and Research, Chennai²

Abstract. This task depends on bootstrap converter with SVM inverter. This work is wanted to improve the speed-guideline of bootstrapped converter-inverter took care of Induction Motor drive (BSCSVM-IIM) using Hysteresis Control. This work oversees assessment among PR-PR and PI-hysteresis controlled BSCSVM-IIM frameworks. The BSC is recommended among rectifier and SVM to upgrade the voltage-pick up .HC is proposed to improve-the-dynamic reaction of BSC-SVM-IIM. The goal of this work is to improve the reaction of BSC-SVM-IIM framework utilizing appropriate regulator in closed-loop. Open loop BSCSVM-IIM framework with disturbance, closed two loop PR-PRC and PI-HC based BSCSVM-IIM frameworks are formed; shown and simulated using Simulink and their results are presented. Assessment is done regarding settling time and SSE. The assessments exhibit the unmatched execution of controlled-BSC-SVM-IIM framework. The proposed framework has inclinations like minimal consonant substance and quick reaction. After reenactment examines, some trial aftereffects of-BSC-SVM-IIM are given to check the adequacy of the proposed converters.

Keywords: BSC-SVM-IIM, SVM inverter, PR-PR and PI-hysteresis, Bootstrap Converter Fed SVM Inverter.

1 Introduction

Presently a-days, research is centered around Multi Level Inverters (MLI's), on the grounds that they can produce voltage waveforms with less contortion than customary inverters dependent on 2level geographies. The upsides of fell MLIs are unmistakable for engine drives and for the utility applications. Due to the incredible interest on medium voltage highpower inverters, the cascaded inverters has gained more interest. As the quantity of levels expands, the integrated yield waveform has more advances delivering a flight of stairs wave that moves toward the ideal waveform

Multiplication of 3phase diode and S.C.R. connect rectifiers for D.C. power -supplies and as front-end-converters for inverter -based- applications, for example, customizable speed -drives and uninterruptible force supply, has brought about genuine symphonious, receptive force, gleam and reverberation issues in modern applications and in transmission/appropriation frameworks. A voltage mutilation because of current-harmonics is a significant issue for utilities. PQissues likewise incorporate uneven and sub-simultaneous recurrence flows which add to voltage hangs and floods, and are the most widely recognized reasons for 'disturbance' stumbling of the movable speed drives.

YongKeun and JongKwang proposed examination of bootstrap circuit activity with a modified PWMdrive plot for a 3phase inverter which is for a brushless dc engine drive[1-2]. A bootstrap circuit for working the high-side protected door bipolar semiconductor of a 3phase inverter for a brushless-dc engine drive was examined hypothetically to maintain a strategic distance from under voltage lockout by limiting the release of a boot-strap capacitor (BSC). Madhusoodhanan proposed a plan and assessment for medium voltage converter relevance's by providing a separated entryway driver power supply. The business gate-drivers were accessible upto 6.5 kV IGBTs. With the advances in the SiC, power gadgets evaluated past 10 kV were being investigated. A medium voltage power converters are used in these gadgets. Business gatedrivers appraised for large- voltages were not accessible. These force gadgets had extremely high dv/dts . High dv/dt acquire difficulties in the entryway driver-design[3]. Yutian explained improved bootstrap strategies for fueling gliding door drivers of flying-capacitor staggered

converters and mixture exchanged capacitor -converters. By utilizing the inalienable properties of staggered converters, these techniques can beat the restriction of traditional boot-strap strategy and make it conceivable to move ground-referred to capacity to the entirety of the coasting -switches[4].

Lei and Liu set a logical technique to assess and plan crossover exchanged capacitor and staggered converters. This explored the utilization of staggered transformation in dc--dc-applications that require an enormous voltage change. A strategy that can fill in as a manual for think about and plan staggered geographies for enormous transformation proportion applications was introduced. The proposed technique kept the conduction misfortune and exchanging misfortune consistent across the various converters[5].

Zhu and Liu focused on the Low-voltage-stress BBC with a high-voltage change pick up. The ordinary BBC had the benefits of straightforward structure, minimal effort, and the capacity to accomplish both voltage venture all over. Be that as it may, because of the negative effects of the parasitic boundaries of the gadget, the voltage change gain of the customary buck-help converter was enormously restricted. Pouladi and Farzanehfard actualized a Single-switch delicate exchanging LED driver appropriate for battery-worked systems[6-8].

Another LED-driver circuit was given dependent on the buck converter coupled inductors for car relevance's. Delicate exchanging condition was accommodated the switch by means of a uninvolved full circuit joining the spillage inductance of the coupled inductors. Sun and Dai proposed Multimode smooth exchanging methodology for dispensing with the operational no man's land in non-reversing buck-boost converter. The presence of operational no man's land during mode exchanging in non-upsetting buck-help converter brings about wavering of the yield voltage and flimsiness of the frame work[9].

The system and impact of the operational dead land were examined, and a multimode smooth exchanging control methodology which can totally dispense with the operational dead land was proposed. Progressed 4mode-adjustment based 4switch non-transforming BBC activity was presented in[10].

Yang explained Analysis demonstrating and execution of an exchanging bi-directional BBC dependent on electric-vehicle cross breed energy stockpiling for V.2.G framework. Battery worked delicate exchanging full BB-LED driver with single attractive component recommended in[11-12]. Another circuit plan of 2switch buck-boost converter was introduced by Jung. A traditional 2switch buck-support (TSBB) converter can work in buck, lift, and buck-boost modes. This

presented another geography for a two switch buck-help converter with similar activity modes. In any case, the proposed TSBB converter had less conductions and exchanging parts than the ordinary TSBB converter, which lessens the force losses[13].

Precise induction of dead land end procedures for the non-transforming coordinated BBC was presented by Zhang. Bidirectional 3-stage DAC converter with implanted DDC converter and transporter based PWM procedure was recommended by Wang. Katherine proposed propels in framework associated PV-power-transformation frameworks. Applications in environmentally friendly power and its control (parc) - corresponding full regulator for semi converter 3phase VSI took care of IMD to upgrade time responses[14-17]. This exertion manages closed-loop semi converter 3 stage acceptance engine drive(SCTPIMD) utilizing PI, FOPID and PRregulator. This exertion proposed PR regulator for SCTPIMD. PR-Fuzzy control improvement of doubly took care of enlistment generator during framework issues was proposed by Mohammad Reza[18-19].

Vector control strategy for IMD dependent on hysteresis regulator and pi regulator near examination was introduced by Shiny[20]. Other technique depended on SVM calculation. Query table DTC technique experienced high waves in force, motion and current just as shifting exchanging recurrence. Then again, SVM based DTC creates equivalently low force, transition and current waves and exchanging recurrence was looked after steady. Tangle at Hysteresis -current control of IMdrives utilizing dSPACE-DSPcontroller[21-22]. Hysteresis current control was moderately a straightforward strategy for PWM procedure with relatively great current circle reaction. This work introduced the equipment usage of hysteresis current control for vector controlled acceptance engine drives utilizing dSPACE-DSP. The examination was centered around the impact of hysteresis data transmission to the engine current quality.

2 Research -Gap

The exceeding effort doesn't include the simulation of PI-hysteresis controlled BSCSVM-IIM frameworks. Hence, this work deal with the simulation of PI-hysteresis controlled BSCSVM-IIM frameworks

4 Simulation Results

The designed converter was first-simulated using MATLAB and then built in the laboratory to authenticate the analysis, design, & enactment of the converter. Open loop BSCSVMIIIM with load disturbance is delineated in Figure 3.

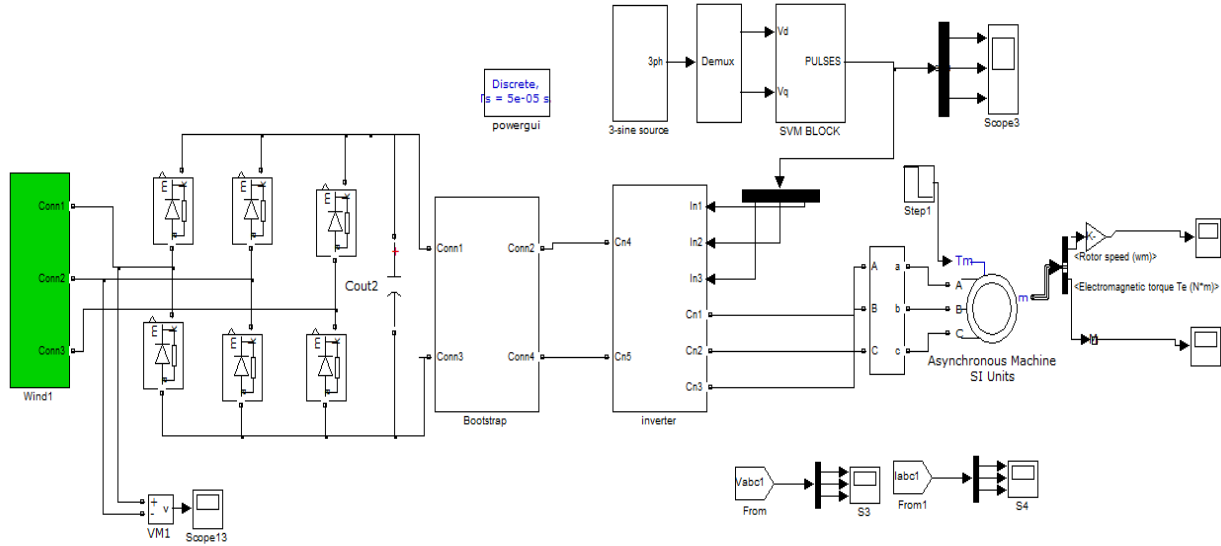


Figure 3: Circuit-diagram -of open loop BSCSVMIIIMwith load disturbance

Input-voltage of BSCSVMIIIM is shown in Figure 4 and the input voltage is 70V. The Circuit diagram of bootstrap converter is outlined in Figure 5.

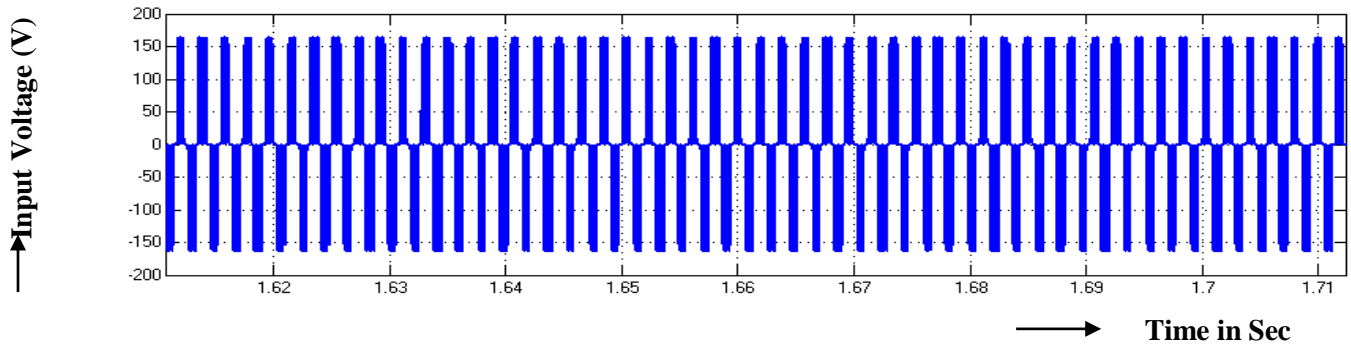


Figure4: Input voltage

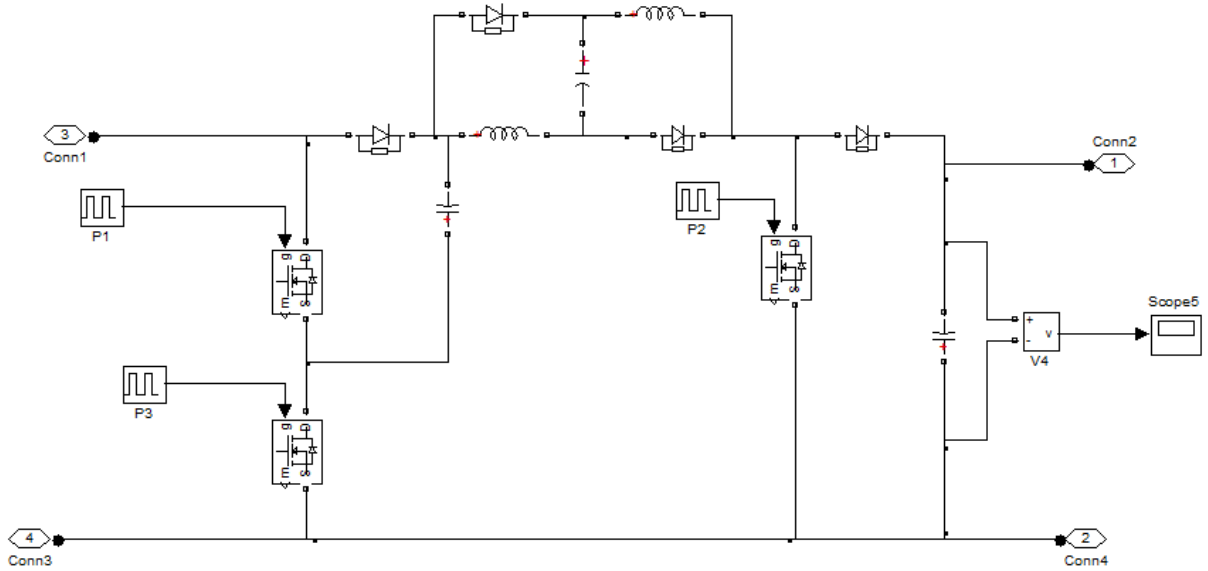


Figure5: Circuit diagram of bootstrap converter

Voltage across bootstrap converter with Bootstrap converter with SVM is presented here. Voltage across bootstrap converter is value is 410 Volts. Voltage across the motor load is 450 Volts and it is indicated in Figure 7.

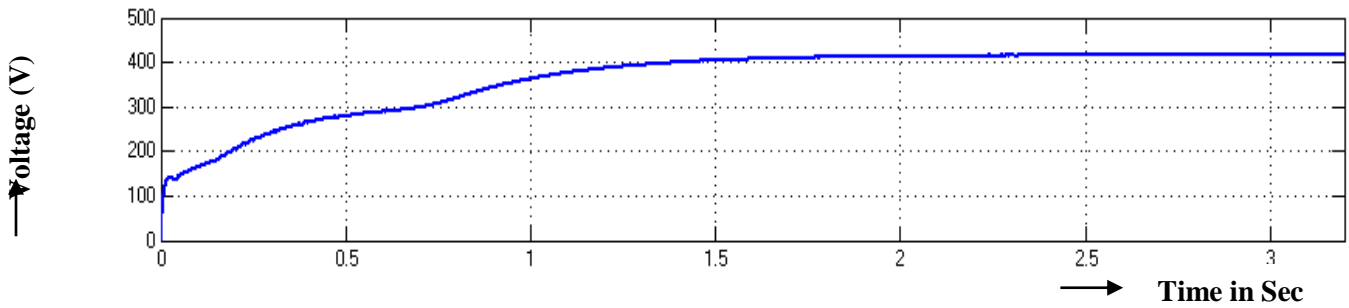


Figure6: Voltage across bootstrap converter

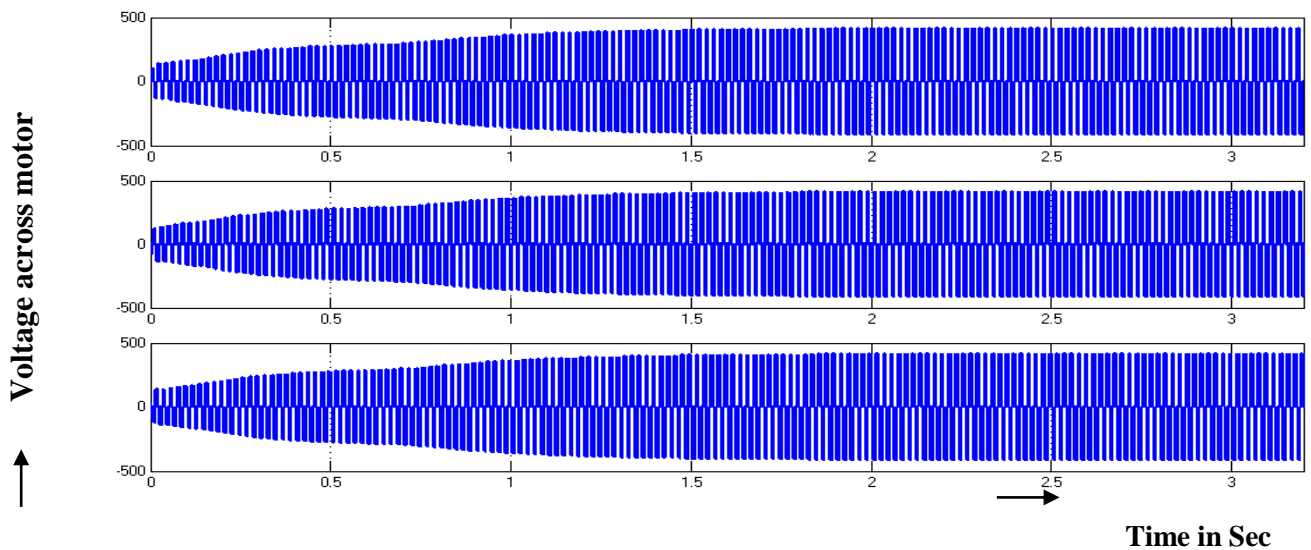


Figure7: Voltage across motor load

Motor speed with Bootstrap converter and SVM is indicated in Figure 8. The value of Motor speed is 1300 RPM.

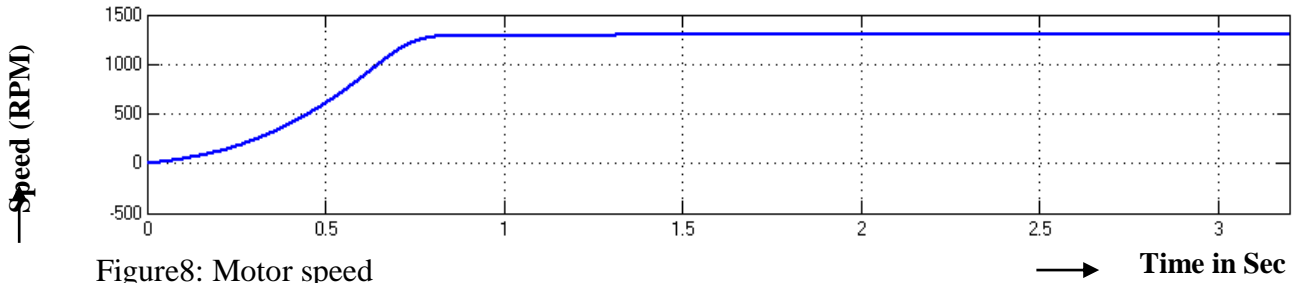


Figure8: Motor speed

The Motor speed of Bootstrap converter with SVM is sketched in Figure 9. The value of Motor speed is 1294 RPM. Motor Torque of Bootstrap converter along with SVM is sketched in Figure 9. The value of Motor Torque is 1 N-m.

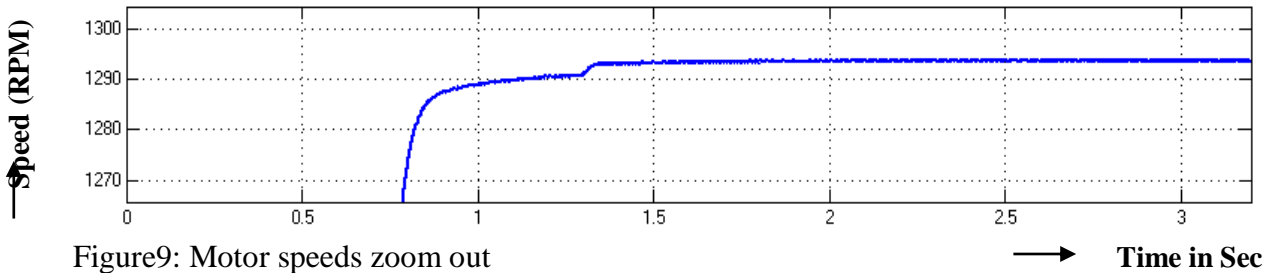


Figure9: Motor speeds zoom out

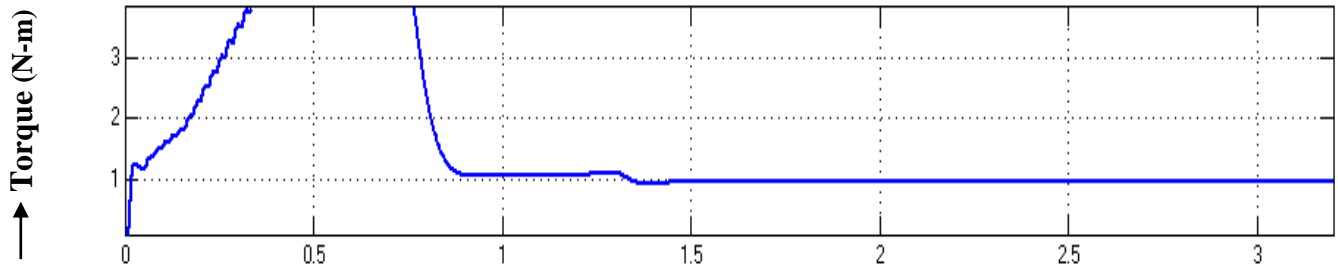


Figure10: Motor Torque

Circuit-diagram of PR-PR controlled closed two loop Bootstrap converter with SVM inverter is appeared in Figure 11.

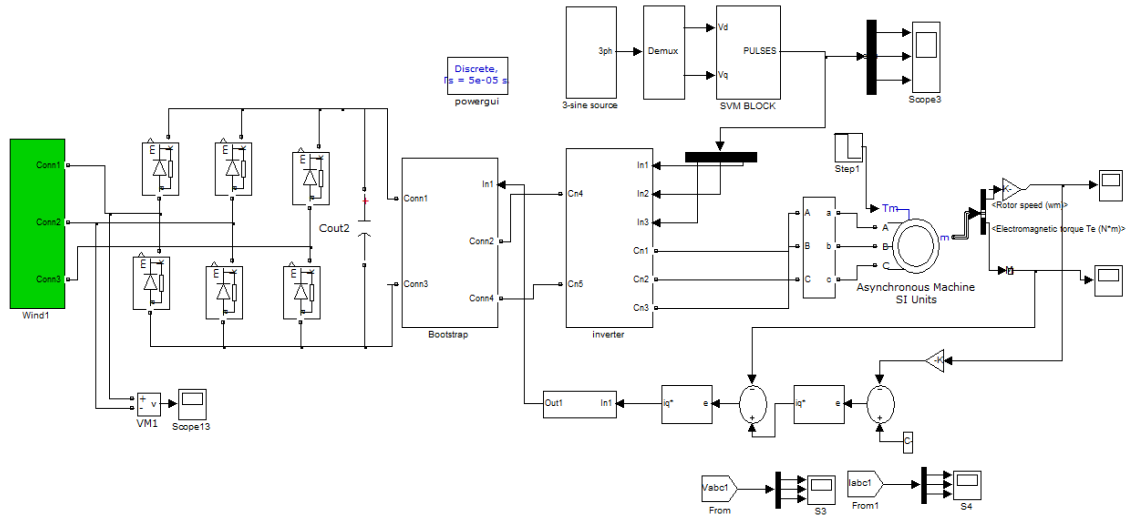


Figure11: Circuit diagram of PR-PR controlled closed two loop Bootstrap converter with SVM inverter

Input voltage of BSC SVM IIM is outlined in Figure 12. The value of input voltage is 175 Volts. Circuit diagram of boot-strap converter is shown in Figure 13.

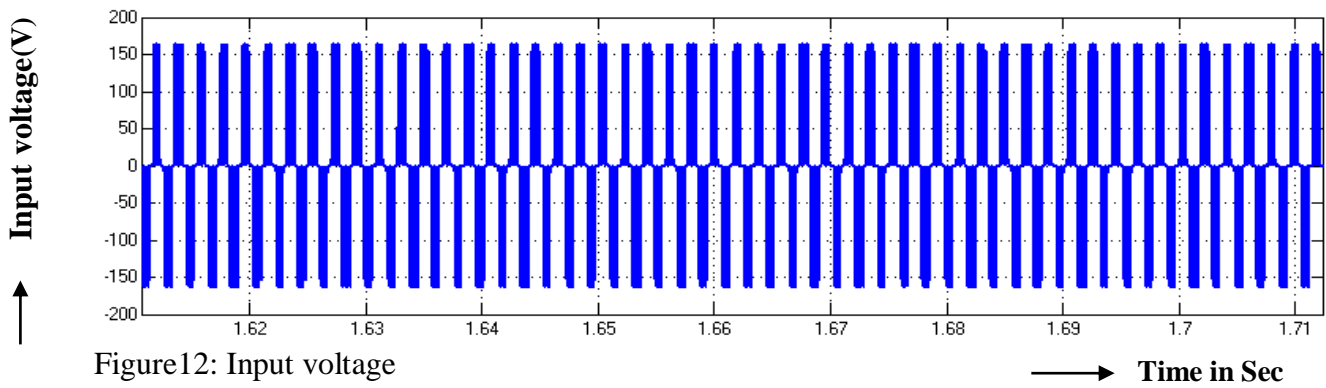


Figure12: Input voltage

Time in Sec

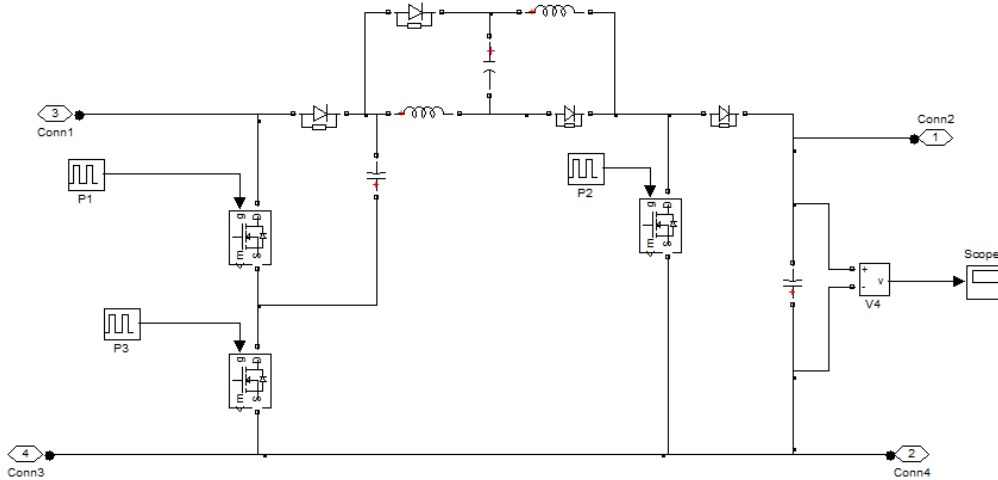


Figure13: Circuitdiagram of bootstrap converter

The Voltage across boot-strap converter and the motor load are figured in Figure 14 and 15. The Voltage across bootstrap converter is 399 Volts and Voltage across motor load value is 400 Volts.

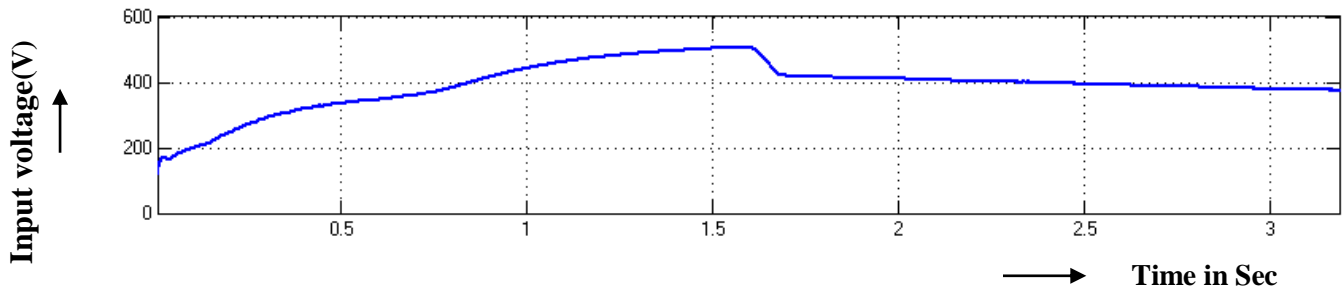


Figure14: Voltage across boot-strap converter

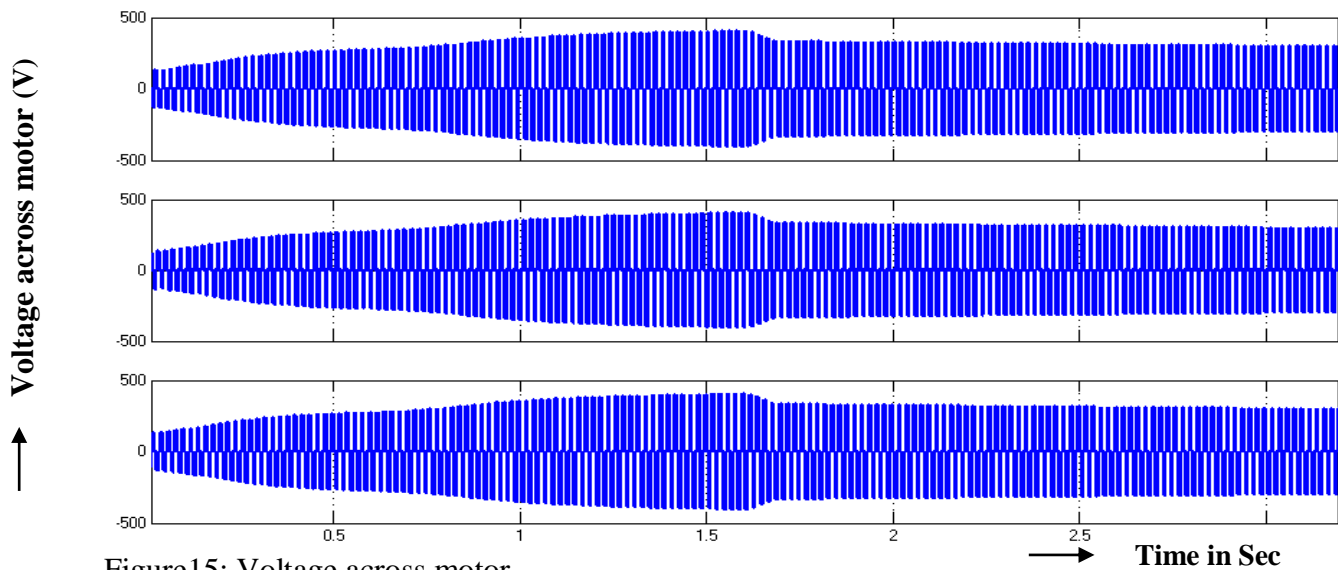


Figure15: Voltage across motor

Motor speed of BSCSVMI-IM is 1450 RPM and is appeared in Figure 16. Motor speed zoom out of BSCSVMI-IM is 1289 RPM and is appeared in Figure 17. The Motor Torque is 0.9 N-m is sketched in Figure 18.

Input voltage of closed two loop BSCSVM IIM with PI-Hysteresis controller is sketched in Figure 20. The value of input voltage is 170 Volts. The Circuit diagram of boot-strap converter is ketched in Figure 21.

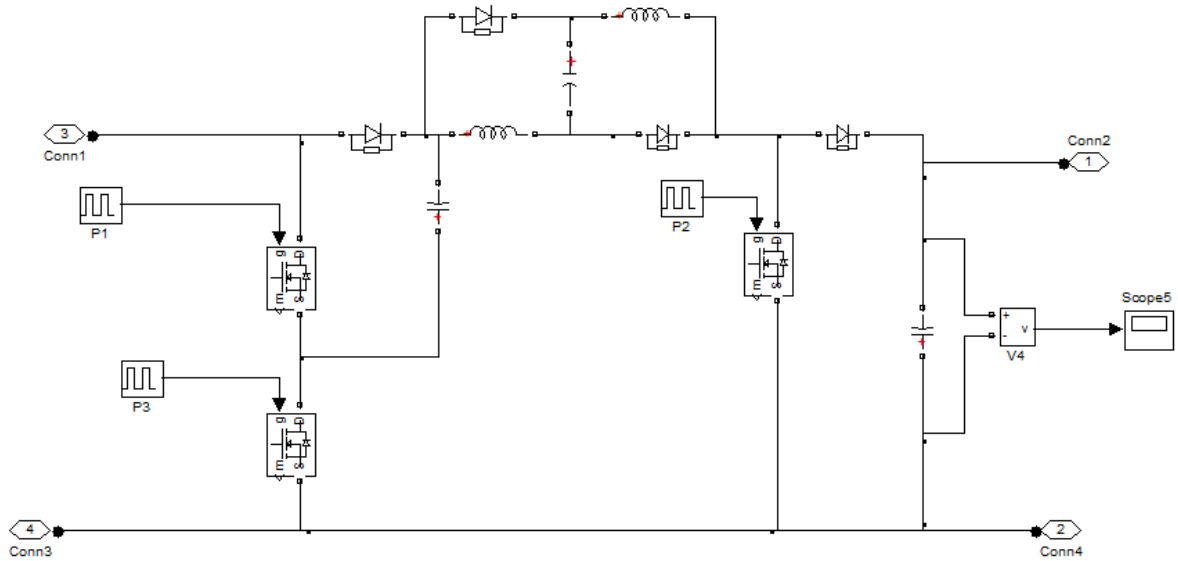
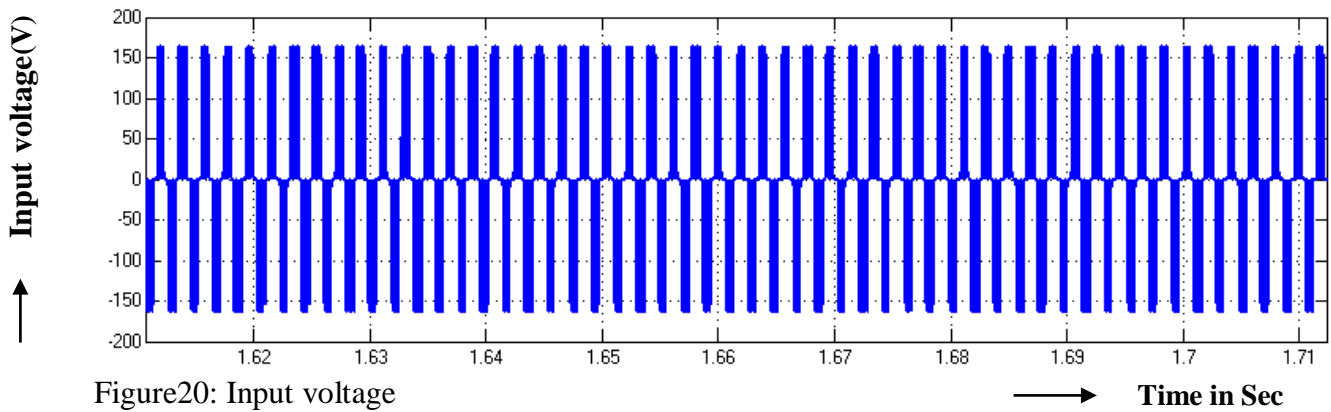
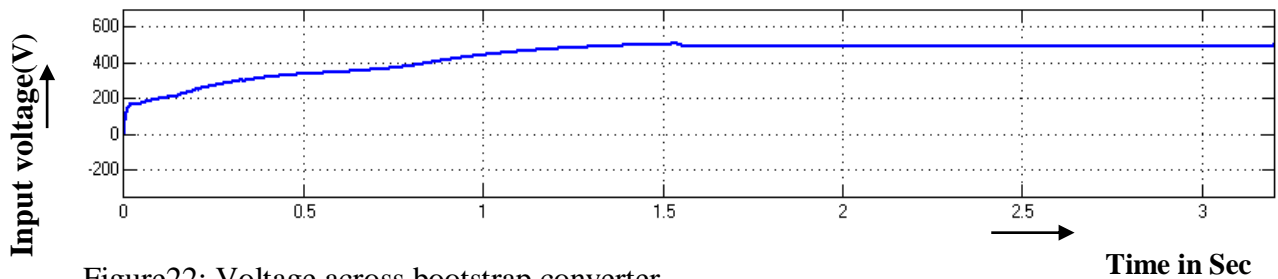


Figure 21: Circuit diagram of bootstrap converter

The Voltage across bootstrap converter and the motor load are indicated in Figure 22 and 23. The Voltage across bootstrap converter is 480 Volts and Voltage across motor load value is 400 Volts.



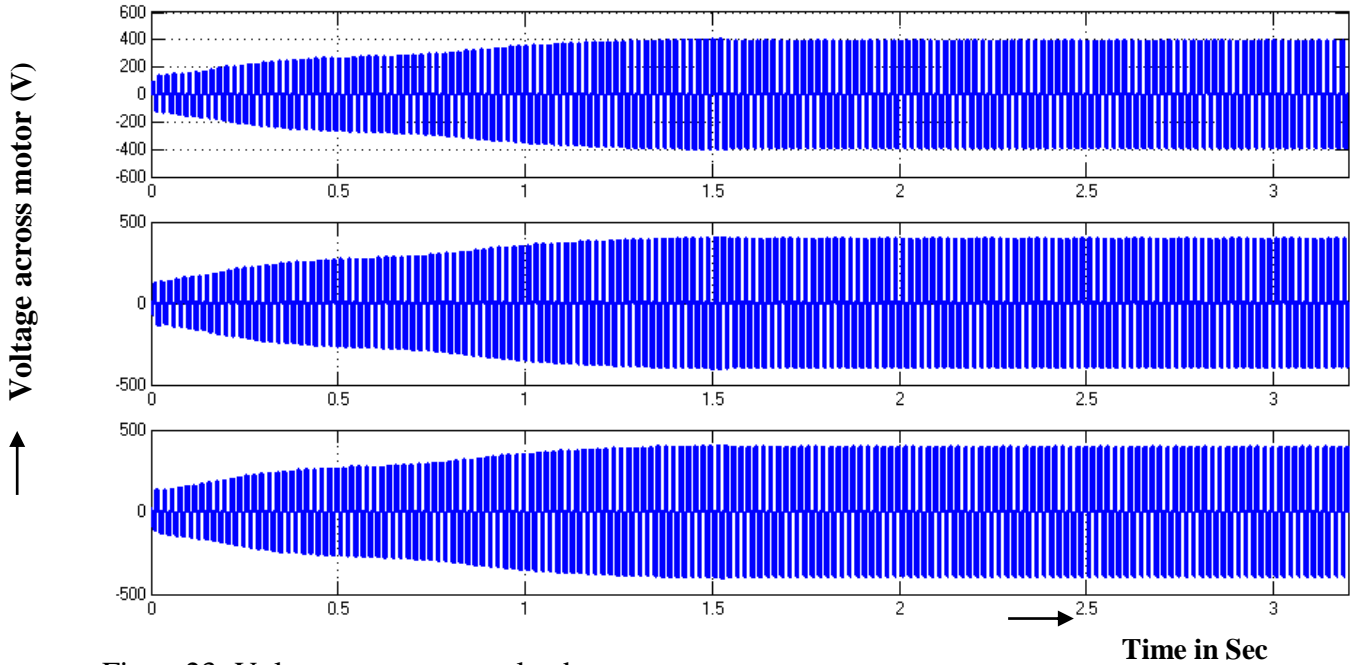


Figure23: Voltage across motor load

Motor speed with SVM inverter of closed two loop FOPID controller is 1290 RPM and it is sketched in Figure 24. Motor speed zoom out along with SVM inverter of closed two loop FOPID controller is 1290 RPM and is sketched in Figure 25. The Motor Torque is sketched in Figure 26.

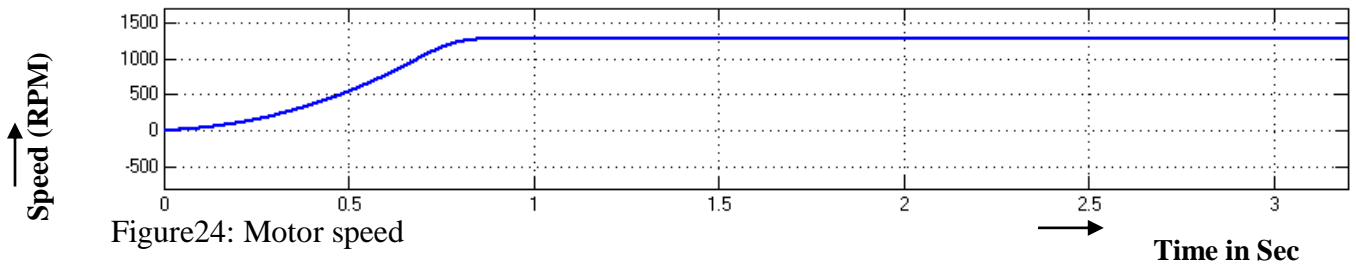


Figure24: Motor speed

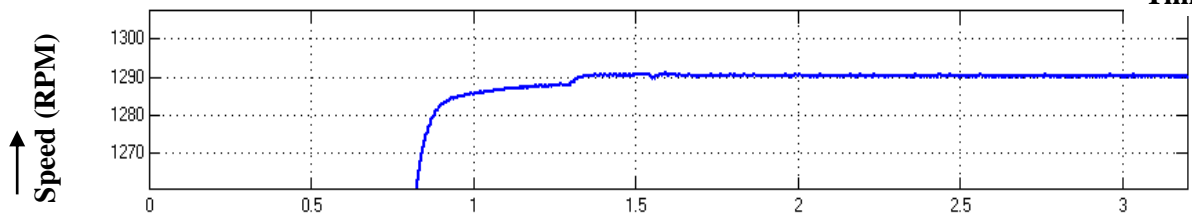


Figure25: Motor speeds zoom out

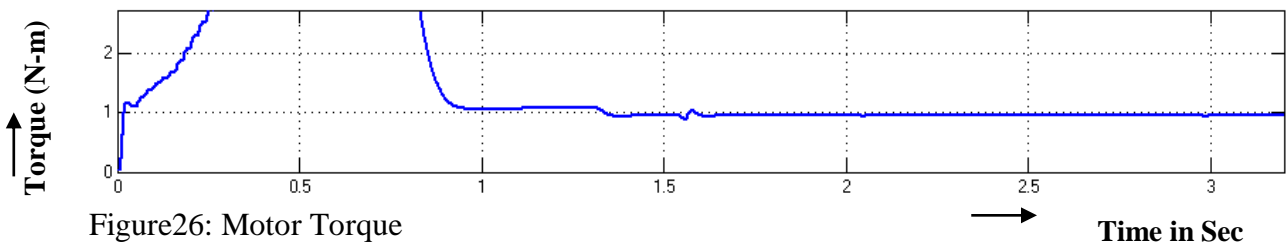


Figure26: Motor Torque

Figure 27 outlines the Bar Chart of motor speed using PIC and HC along with its various Time Domain Parameters. The table-1 explains the comparative relation of Time Domain Parameters for motor-speed using PIC-HC and PR-PR, the 'Tr (rise-time)' is lessened from 1.34

Sec to 1.32 Sec; ‘Ts (Settling-time)’ is lessened from 1.75 Sec to 1.60 Sec; ‘Tp (peak-time)’ is lessened from 1.55 Sec to 1.47 Sec; ‘Ess (Steady-state error)’ is lessened from 0.5 RPM to 0.3 RPM.

Table-1

The comparative relation of Time Domain Parameters for motor speed using PI-HC and PR-PR

Controller	Tr(S)	Ts(S)	Tp(S)	Ess(RPM)
PR-PR	1.34	1.75	1.55	0.5
PI-HC	1.32	1.60	1.47	0.3

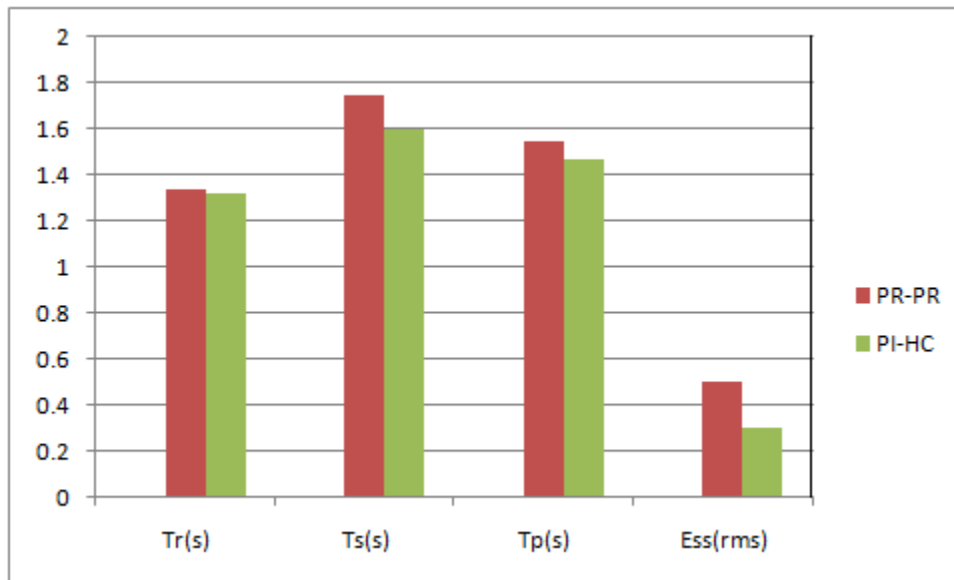


Figure 27. Bar Chart of Time Domain Parameters (Tr, Ts, Tp, Ess) for motor speed using PIC-HC and PR-PR

Figure 28 outlines the Bar Chart of motor torque using PIC-HC and PR-PR along with its various Time Domain Parameters. The table-2 explains the comparative relation of Time Domain Parameters for motor-torque using PIC-HC and PR-PR. By using PIC-HC, the ‘Tr (rise-time)’ is lessened from 1.35 Sec to 1.33 Sec; ‘Ts (Settling-time)’ is lessened from 1.82 Sec to 1.67 Sec; ‘Tp (peak-time)’ is lessened from 1.71 Sec to 1.67 Sec; ‘Ess (Steady-state error)’ is lessened from 0.3 N-m to 0.1 N-m.

Table-2

The comparative relation of Time Domain Parameters for motor torque using PI-HC and PR-PR

Controller	Tr(s)	Ts(s)	Tp(s)	Ess(N-m)
PR-PR	1.35	1.82	1.71	0.3
PI-HC	1.33	1.67	1.67	0.1

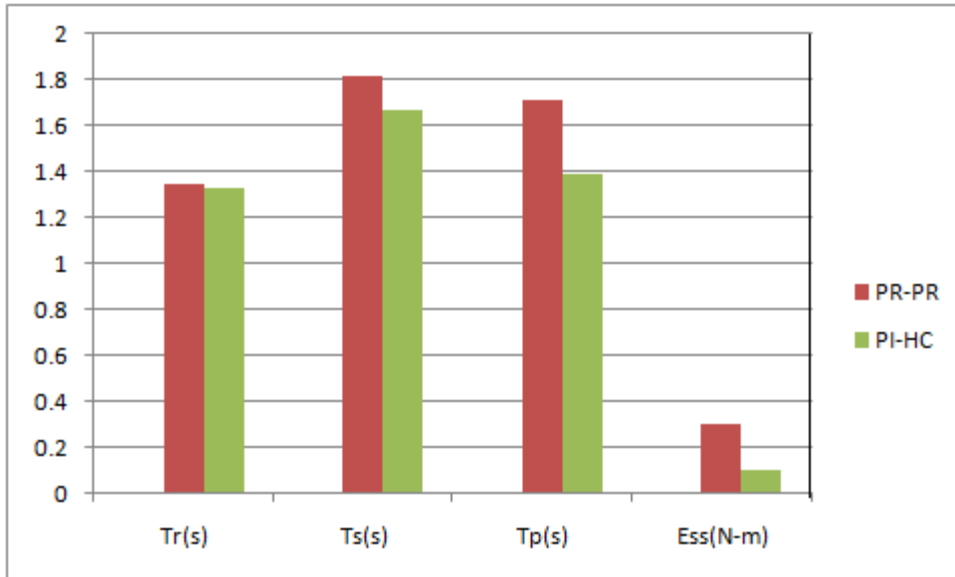


Figure 28. Bar Chart of Time Domain Parameters (Tr, Ts, Tp, Ess) for motor Torque using PI-HC and PR-PR

5 Experimental Results

Hard-ware snap-shot of IM based Boot-strap converter with SVM inverter system is sketched in Figure 29. The hardware of BCSVMI consists of Motor load, inverter panel, BCB, control panel and RBC panel. The snap shot of the input voltage is outlined in Figure 30.

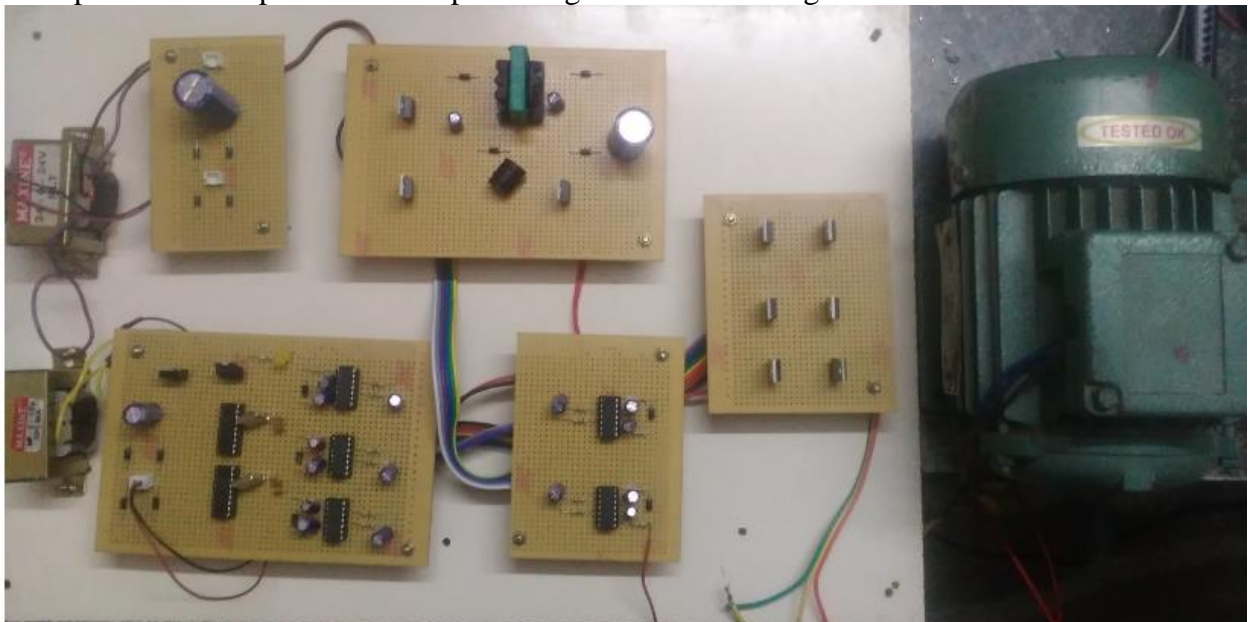


Figure 29: Hardware snap shot

The Switching pulse for bootstrap converter across S1 and S2 is appeared in Figure 30.

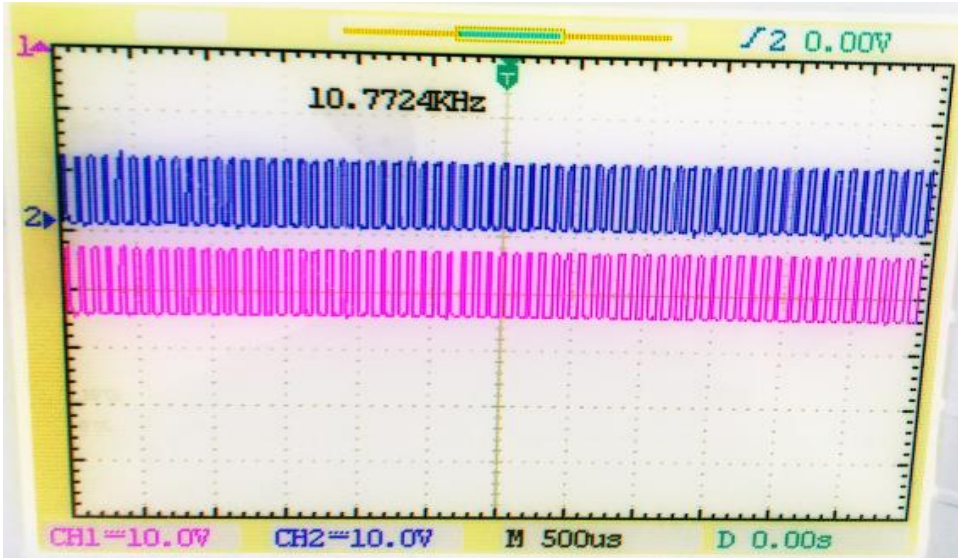


Figure 30: The Switching pulse for bootstrap converter across S1 and S2, Switching-pulse for inverter M1, M2 and M3, M6 are outlined in Figure 31 and Figure 32 respectively. From Figure 31 it is clear that its peak to peak value is increased.

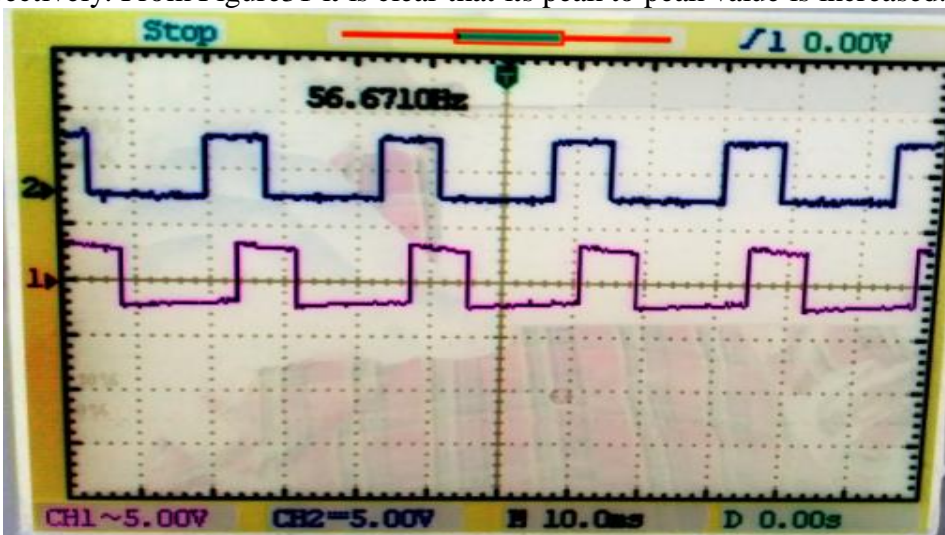


Figure 31: Switching pulse for inverter M1, M2

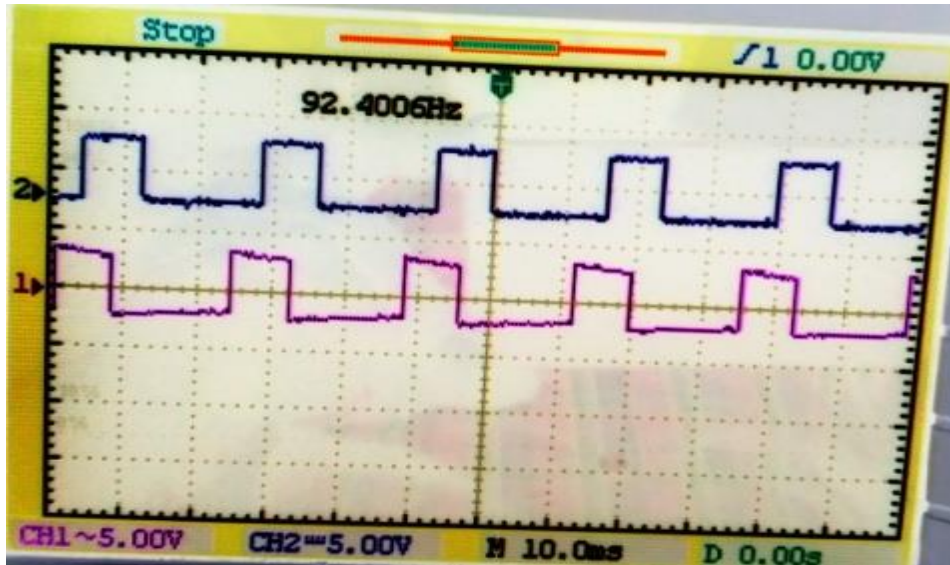


Figure32: Switching pulse for inverter M3,M6

The Motor load voltage and the motor current are outlined in Figure 33 and 34 respectively.

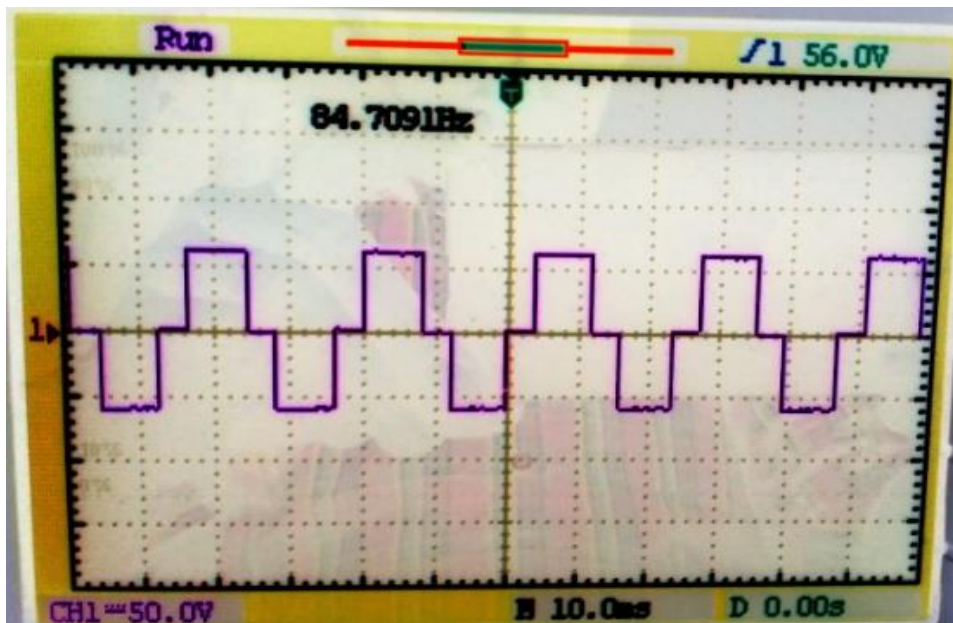


Figure33: The motor load voltage

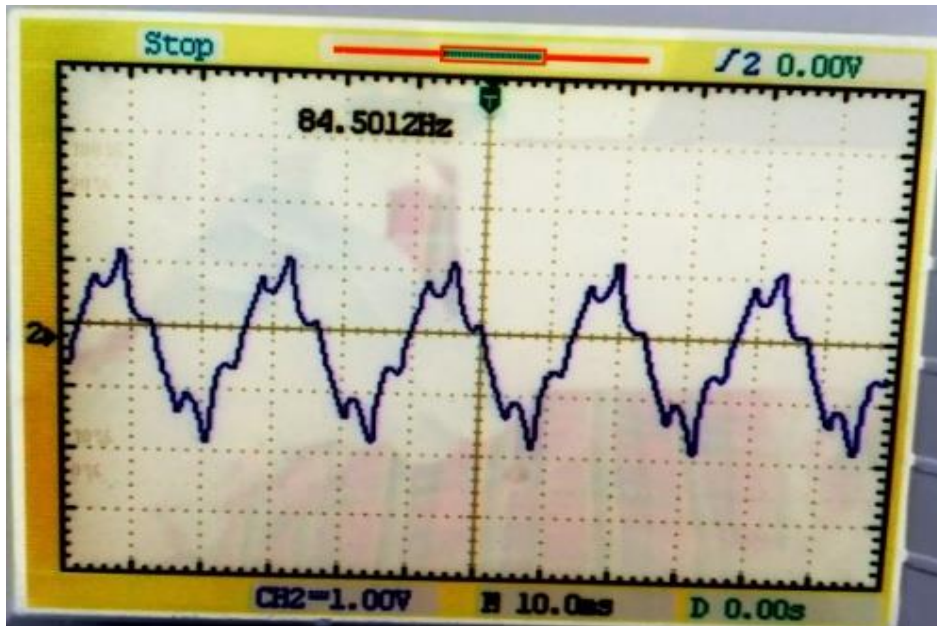


Figure34: The motor load current

6 Conclusion

The Closed two loop BSC- SVM systems along with PR-PR and PI-HC are designed and simulated using the simulation tool Mat-labSimulink. The outcomes of the simulations for the closed two loop systems with PR-PR and PI-HC are tabulated. The time domain parameters are obtained and the values are compared and represented using bar chart. The T_s (settling time) is reduced to 1.67 Sec and E_{ss} (steady state error) is reduced to 0.1N-m by using HC (Hysteresis controller). From the obtained response of outcome of PI-Hysteresis Controller system is better than the PR-PR controlled system. The hard-ware of BSC-SVMI is fabricated and tested. The experimental results of SVM have been presented for validation purpose. The proposed system has few benefits like fast response and low harmonics. The downside of bootstrap converter is that it is appropriate for low power. The present work brings out the performance of PI-hysteresis controlled BSC-SVMIIM. The investigations on Slide Mode controlled BSC-SVMIIM will be done in future.

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