## Machine-Converter Voltage Source for Autonomous Systems of Distributed Power Supply

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## Abstract

The paper presents electric Machine-Converter Voltage Source (MCVS) for systems of distributed power supply with shaping output voltage in an electric machine (generator). The MCVS consists of electric machine and a cycloconverter that has three-phase bridge circuit. Both synchronous generator and asynchronous exciter, which are located in one housing, operate as electric modulation generator. Functional diagram of complex MCVS, where the output voltage is stabilized by simultaneous or separate use of several control channels. We analyzed output voltages of MCVS and dependences of non-sinusoidality and distortion factors on ratio rotational speed of generator to output voltage frequency of MCVS for various amplitude modulation indexes and control modes of cycloconverter.

**Keywords:** Machine-converter voltage source, autonomous distributed power supply system, output voltage envelope, cycloconverter, harmonic analysis, distortion factor, control optimization.

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### 1. Introduction

One of the main goals of designing Machine-Converter Voltage Sources (MCVS) for autonomous power supply systems is outputting electric power, which parameters meet the quality standards [1–7, 11]. Main quality factors of independent power supply sources are rated magnitude and frequency of output voltage. Deviations of the load, supply voltage and currents are reasons of failures of expensive equipment. Therefore, research and development circuits of Machine-Converter Voltage Sources (MCVS) as main component of autonomous distributed power supply systems that can stabilize output parameters is a problem of current importance.

# 2. Autonomous sources with shaping output voltage waveform and stabilization of electric output parameters in electric machine

We analyzed existing stabilization methods [10–12, 14] of output electric parameters (output voltage magnitude and frequency) of systems «electric generator – static frequency converter». The analysis shows that to generate sine wave of output voltage by means of electric generator (EG) the most viable method for wide-range variation of the generator speed and the power consumed is voltage regulation in the generator excitation windings [9]. Besides, the three-phase voltage system of stable frequency can be generated in an autonomous source with electromagnetic shaping of the output voltage waveform, which consists of two electric AC machines that are mount on the same shaft by recovering modulating functions by means of static converter.

Figure 1 shows functional diagram of autonomous singlephase voltage source of stable frequency.



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Figure.1. Functional diagram of autonomous single-phase voltage source of stable frequency with shaping the output voltage waveform of electric generator

Here modulation type electric generator (EG) consists of synchronous generator (SG) and asynchronous exciter (AE), which are mounted on the same shaft and located in one housing [8, 12]. Windings of the electric machines of EG are made with different number of pole pairs, which allows providing independence of output voltage frequency on rotation speed of SG shaft. Excitation winding of SG is connected to the rotor winding of AE via controlled rectifier (CR). Stator winding of AE is connected to autonomous voltage inverter, and the armature winding of SG – to cycloconverter (CC). An additional three-phase winding, which is connected to rotor winding of AE, is wound on rotor of SG. Output armature winding of SG is carried out as three single-phase winding, which are shifted in space for angle  $\frac{2\pi}{3}$  el. degrees. EG outputs power to the load via

three-to-one-phase cycloconverter (CC).

Operation of MCVS with shaping the output voltage waveform in electric machine can be explained using figure 1. If the rotor is rotated and the winding of AE is supplied by PWM autonomous inverter, an EMF is induced in rotor winding of AE with frequency given by:

$$\omega_2 = \omega_1 p_1 + \omega_0, \qquad (1)$$

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where  $\omega_1$  is angular frequency of rotational EMF of generator;  $p_1$  is pole pair number of windings of SG;  $\omega_0$  is angular frequency of stator winding current of AE.

Excitation winding of SG is a load for asynchronous exciter via controlled rectifier. In armature winding of synchronous generator an EMF is induced by DC excitation winding given by:

$$e_{1}(t) = E_{1m} \sin(\omega_{1} p_{1} t)$$

$$e_{1}(t) = E_{1m} \sin(\omega_{1} p_{1} t) \frac{2\pi}{3}$$

$$e_{1}(t) = E_{1m} \sin(\omega_{1} p_{1} t) \frac{2\pi}{3}$$
(2)

Angular frequency of the induced EMF is given by:

$$\omega_4' = \omega_1 p_1.$$

Additionally wound three-phase winding of SG is also the load for AE. Phase-to-neutral EMFs of armature windings of SG that are caused by the field of additional three-phase winding of SG are functions of angular frequency  $\omega_4''$  in terms of formulae:

$$e_{2}(t) = E_{2m} \sin\left[\left(\omega_{1}(p_{2}+p_{3})+ _{0}\right)t\right]$$

$$e_{2}(t) = E_{2m} \sin\left[\left(\omega_{1}(p_{2}+p_{3})+ _{0}\right)t-\frac{2\pi}{3}\right],$$

$$e_{2}(t) = E_{2m} \sin\left[\left(\omega_{1}(p_{2}+p_{3})+ _{0}\right)t+\frac{2\pi}{3}\right],$$

$$w_4'' - w_4' = \omega_1 (p_2 + p_3) + \omega_0.$$
(3)

where  $p_2$  is pole pair number of windings of AE;  $p_3$  is pole pair number of additional three-phase excitation winding of SG.

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Adding together the two EMFs in armature winding of SG the resulting EMF is generated, which magnitude is oscillating with beat frequency given by:

$$e_{A}(t) = 2E_{m} \sin\left[\frac{\left(\omega_{1}\left(p_{1}+p_{2}+p_{3}\right)+\omega_{0}\right)}{2}t\right] \cos\left(\frac{\omega_{0}}{2}t\right)$$

$$e_{B}(t) = 2E \sin\left[\frac{\left(\omega_{1}\left(p_{1}+p_{2}+p_{3}\right)+\omega_{0}\right)}{2}t - \frac{2\pi}{3}\cos\left(\frac{\omega_{0}}{2}t\right)\right]$$

$$e_{C}(t) = 2E \sin\left[\frac{\left(\omega_{1}\left(p_{1}+p_{2}+p_{3}\right)+\omega_{0}\right)}{2}t - \frac{2\pi}{3}\cos\left(\frac{\omega_{0}}{2}t\right)\right]$$

$$\omega_4'' - \omega_4' = \omega_1 \left( p_1 - p_2 - p_3 \right) + \omega_0.$$
 (4)



As result of extraction of the output voltage envelope by means of cycloconverter CC, we got independence of beat frequency  $\omega_4$  on the generator shaft speed, i.e.

$$\omega_4 = \omega_0 \,. \tag{5}$$

But here it is necessary to choose number of pole pairs of windings of synchronous generator, asynchronous exciter

and additional three-phase excitation winding of SG in terms of formula:

$$p_1 = p_2 + p_3. (6)$$

In figure 2 and 3 you can see graphic illustration of shaping the EMF waveforms of armature windings of SG and output voltage envelope of autonomous MCVS.



Figure. 2. Waveforms of induced EMFs and resulting EMFs of phases of armature winding of SG



Figure. 3. Waveforms of EMFs of armature windings of SG and output voltage envelope



## 3. Combination circuit of Machine-Converter Voltage Source with stabilization of output parameters

Stabilization of output voltage of MCVS with shaping output voltage waveform in electric generator can be realized via several channels: by means of DC excitation winding circuit; by means of AC excitation winding circuit; by means of cycloconverter circuit. Furthermore, one is able to use simultaneously two or all of the channels enabling fine varying amplitude modulation percentage and stabilizing output electrical parameters.

In figure 4 a functional diagram of MCVS circuit is shown, in which the output voltage can be stabilized by means of several control channels.



Figure 4. Functional diagram of MCVS of stabilizing system of the electrical output voltage parameters of combination type

In MCVS circuit, which uses stabilization of the output voltage by means of DC excitation winding circuit [2, 10, 12], control system (CS) of autonomous voltage inverter (AVI) with PWM (see figure 4), has constant reference signal of frequency and magnitude of excitation current of AE, which does not depend on output voltage and frequency of MCS. Input signals of the Electrical Generator Control System (EGCS) are: speed of driving shaft  $\omega$  and reference value of voltage  $U_{REFU}$ , which is compared with current value of output voltage amplitude (in Difference Block (DB)). Regulation of magnitude of output voltage of MCVS is carried out by changing angle  $\alpha$  of Controlled Rectifier (CR), which feeds the circuit of DC excitation winding of generator EG.

In MCVS circuit, which realizes stabilization of the output voltage by means of AC excitation winding circuit [2, 10, 12] the output voltage is regulated by means of autonomous voltage inverter AVI, which feeds the stator circuit of AE. In addition, in DC excitation winding circuit of SG can be applied Controlled Rectifier (CR), instead of the uncontrolled one. However, it leads to certain limitations for practical use of such MCVS, since with no ability to correct the biasing current of SG the effective range of the output voltage stabilization will be reduced, when the angular speed of SG is changed [13, 14]. CC has its own control system with firing angles ( $\alpha$ ) and demodulates

output voltage of SG. The firing angle of cycloconverter CC is generated subject to signals from current sensor (CS) and voltage sensor (VS) [9, 12, 14].

Joint use of these stabilization methods in circuit of MCVS, shown in fig. 4, allows increasing control accuracy and expanding speed range of driving shaft even for one-channel excitation system of generator EG.

In such a way the three-phase voltage system of stable frequency can be achieved in MCVS with shaping the output voltage waveform, which is made of two electric AC machines that are mount on the same shaft, and then by extracting modulating function by means of cycloconverter CC. This enables to get an output voltage of MCVS the one-phase voltage, which is close in waveform to sine wave, with frequency that is equal to half of AC excitation current frequency of AE and does not depend on the rotational speed of EG [10, 15].

Main advantage of MCVS with shaping the output voltage waveform in electric machine is overlapping electric machines with different number of pole pairs on one magnetic core, which allow providing no sliding contacts and increasing reliability of the system. Introduction of additional components that operate in feedback circuit carry out excitation of generator, stabilization of magnitude and frequency of the output voltage via two channels: control of excitation and of the converter circuit. Ability to control the amplitude modulation percentage in wide range allows achieving good power factors of MCVS with high quality of the output voltage waveform.

After researching MCVS [8, 9, 11, 12, 15] we have found out that the waveform of the output voltage envelope depends on amplitude modulation percentage index and decreasing the index causes «delaying» of the envelope in regions with small voltages, which leads to deviation of the output voltage waveform from sine waveform. One can solve this problem most effective due to optimization of the control mode of CC [10, 12, 17].

Introduction of additional components into feedback of CC and using closed loop control system by introducing the voltage feedback allows improving quality of output voltage [10, 15-17].

## 4. Harmonic analysis of the load voltage of autonomous Machine-Converter Voltage Source with stable frequency

For comparative analysis of studied load voltages of MCVS for various control modes of CC we have to determine non-sinusoidality factor  $k_n$ , distortion factor  $k_U$  and their dependence on ratio of rotational speed of generator to output voltage frequency  $\frac{\omega_{rot}}{\omega_{out}}$  for various

amplitude modulation indexes k.

These factors are determined in terms of formulae:

$$k_n = \frac{U_{\Sigma}}{U_{H1}}, \ k_U = \frac{U_{H1}}{U_H},$$
 (7)



where  $U_H$  is effective value of output voltage;  $U_{H1}$  is effective value of fundamental harmonic of output voltage;  $U_{\Sigma}$  is total effective value of higher harmonics of output voltage.

In diagrams (figures 5–8) you can see functional dependences of the factors on ratio of rotational speed of generator to output voltage frequency of MCVS for various amplitude modulation indexes.



Figure 5. Functional dependence of non-sinusoidality factor of output voltage on ratio of rotational speed of generator to output voltage frequency of MCVS with open control system of CC



Figure 6. Functional dependence of non-sinusoidality factor of output voltage on ratio of rotational speed of generator to output voltage frequency of MCVS with closed control system of CC



Figure 7. Functional dependence of distortion factor of output voltage on ratio of rotational speed of generator to output voltage frequency of MCVS with open control system of CC



Figure 8. Functional dependence of distortion factor of output voltage on ratio of rotational speed of generator to output voltage frequency of MCVS with closed control system of CC

From graphs shown in figures 5–8 you can see that harmonic structure of MCVS voltage, when changing load, is distinctly better due to optimization of the control mode of CC in the range up to double ratio of rotational speed of generator to output voltage frequency. Particularly at the double ratio the non-sinusoidality factor does not exceed standard value (5%) for k = 0,8.

## Conclusions

Autonomous sources with shaping the output voltage waveform at output of EG by adding EMFs of close frequencies are prospective and provide high quality of output voltage waveform in autonomous systems of distributed power supply.

We found that the machine-converter source described in the paper provides output voltage control in required range, as well as it maintains the frequency and magnitude of output voltage in accordance with requirements to autonomous power supply systems.

We discovered that due to optimization of the control mode of CC the quality of output voltage could be improved. Particularly at the double ratio the non-sinusoidality factor does not exceed standard value (5%) for k = 0.8.

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