Reliability Considerations in Mobile Devices

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ABSTRACT

The problem of reliability in current chips has been the subject of numerous researchers. Mobile devices, commonly used in multimedia communications require low power during both normal operation and testing. In this paper a novel algorithm is presented for embedding test sets containing don’t care values into sequences generated by binary counters. Therefore, both test time and power consumed during testing of the chips can be considerably reduced.

Keywords
Low power testing, Mobile device reliability.

1. INTRODUCTION

The reliability problem in current electronic devices is the subject of ongoing research [1]. Mobile devices, apart from the ever-existing need for low power during normal operation suffer from an increasing demand for low power during testing. Various schemes have been proposed in order to provide efficient low-power testing capabilities. One class of proposed solutions, namely Built-In Self Test (BIST) schemes has been shown to be irreplaceable for testing mobile devices; BIST schemes utilize on-chip circuitry to generate test patterns and verify the respective responses of the circuit. Therefore, they drive down the cost of test and boost the quality of the resulting testing scheme possessing various advantages, e.g. low cost, high quality of the delivered IC for both modeled and non-modeled faults and the possibility for performing at-speed test. It is for these (and other) reasons that BIST schemes seem to have been well accepted by the IC design and manufacturing community [2].

The BIST circuitry of a combinational CUT comprises a test generator (TG) that generates the required test patterns (and applies them to the CUT inputs) and a Response Verifier (RV) that compresses the CUT responses and evaluates the compressed output. Pseudorandom BIST schemes utilize sequences generated by modules that can be found on-chip (e.g. counters or accumulators) or can be easily implemented by modifying existing registers (e.g. Linear Feedback Shift Registers, LFSRs [3]). The main advantages of the pseudorandom testing paradigm include the low hardware overhead and the simplicity of the BIST control. Counters have been successfully utilized as BIST pattern generators in the BIST context [5], [8], since these modules commonly exist into current devices.

Among the pseudorandom schemes that have been proposed, the test set embedding paradigm that tries to embed a given test set into the sequence generated by a pseudorandom generator has gained attention by researchers [4], [5], [6], [7]. A test set embedding scheme typically utilizes a test set T and a hardware generator G and tries to embed T into the sequence generated by G, in such a way that the length of the selected subsequence is minimized. The advantage of test set embedding stems from the fact that the pseudorandom generator used has moderate or low hardware overhead.

Firstly, Lempel et al, utilized theory of discrete logarithms to embed test patterns into LFSR-generated sequences [4]. Kagaris and Tragoudas utilized counters, and by using negation and permutation of the counter outputs succeeded to generate complete test sets into the thereof generated sequences within acceptable time limits [5], [8]. Dorsch [6] proposed an algorithm for embedding test patterns, and, as a next step, whole test sets into sequences generated by accumulators. In a recent work [9], an algorithm was presented that, given an n-stage accumulator accumulating a constant pattern B, calculates in O(n) time, the location of any given pattern V into the generated sequence.

In this paper we propose a novel algorithm that, given a test set containing don’t care (X) values, calculates a minimum sequence generated by a binary counter that generates all the test patterns of the test set. Therefore, the power generated for the testing of the module is deteriorated.

In order to provide the framework in which the proposed algorithm operates, we shall present the typical procedure one would follow in order to generate a test set for a given circuit. At first, the description of the CUT is entered into an Automatic Test Pattern Generation (ATPG) tool e.g. [12]. The ATPG tool generates test patterns (usually containing don’t care values). Next, the test set is compressed, in an attempt to decrease the number of patterns. In effect, some of the ‘X’ values are replaced with 1’s or 0’s (some times arbitrarily) without necessarily increasing the fault coverage. The output of the tool is thus a test set containing 0’s and 1’s.

The above-mentioned procedure is viewed differently from the BIST designer viewpoint. BIST engineers, especially those targeting the problem of embedding test patterns, firstly simulate
pseudorandom patterns in portions of a predetermined number e.g., 32K patterns. When the simulation comes to a point where a few portions (e.g. 3 consecutive applications of 32K patterns) do not detect any new faults, the remaining faults are labeled as 'non-randomly testable' or so-called 'hard-to-detect' faults [8]. It is for these faults that test patterns (containing don't cares) are generated and test embedding is applied. The rationale of this approach is that if hard-to-detect faults are detected with a pseudorandom sequence of adequate length, then pseudorandomly testable faults will be also detected. Experimental results indicate that this approach gives fruitful results, e.g. in [4], [5], [6], [8].

The exploitation of don’t care values drives down the number of required patterns as is shown experimentally in the sequel.

In this paper we present a novel algorithm for Embedding Test sets containing don’t care values (TEX) into counter-generated sequences that, given a test set containing vectors with don’t care bits, calculates an optimal subsequence of the sequence generated by a binary counter that contains the required test set.

The paper is organized as follows. In Section 2 some definitions and Lemmas are provided, that are utilized in Section 3 where the proposed algorithm is presented, exemplified and analyzed; in Section 4 experimental results of the application of the proposed algorithm into some test sets are presented. We conclude the paper in Section 5.

2. DEFINITIONS, NOTATIONS AND LEMMAS

We denote by L(V) the location of an n-bit pattern V in the sequence generated by an n-stage counter starting from the all-zero value, i.e. the number of cycles that the counter needs to operate in order to generate V. For example, for n=3, the location of V=6 in the generated sequence is L(6) = 6. Furthermore, the distance of two patterns V1 and V2 in the sequence generated by a counter is the difference L(V1) - L(V2).

Definition 1: Consider an n-bit vector V containing k don’t care bits, 0≤k≤n. The expanded list of V is a list of 2^k vectors, each one of consisting only of ‘0’ and ‘1’, in such a way that all 2^k combinations of ‘0’ and ‘1’ appear into all positions held by ‘X’ in the vector V.

From Definition 1, if V has k don’t care bits, then there are exactly 2^k vectors in its expanded list. For example, let n = 5 and V=01X1X. Then k = 2 and the expanded list of V consists of the 2^k = 4 binary vectors: {01010, 01011, 01111, 11111}.

Definition 2: Consider a test set T comprising t test vectors containing don’t care values and the expanded lists L1, L2, ..., Lt of the t test vectors V1, V2, ..., Vt respectively. Let Ti be a vector belonging to one (or more) of the expanded lists of Vt where 1≤i≤t. The including list of a vector Ti in the test set is a list of the including lists this vector belongs to.

For example, let us consider the test set consisting of 9 5-bit test vectors, given in Figure 1(a). The expanded lists of the test vectors of the test set are presented in Figure 1(b). The decimal notation of each vector is presented in Figure 1(c) for reference purposes. Then, the including list of V=3 is {L7, L9}, while the including list of V=19 is {L3, L7, L9}. In Figure 2 the including lists of all the test vectors of the test set presented in Figure 1 are presented.

\[
\begin{align*}
V1 &= 101XX \\
V2 &= X10XX \\
V3 &= 100XX \\
V4 &= 0111X \\
V5 &= X10X0 \\
V6 &= X00X0 \\
V7 &= XX011 \\
V8 &= XX101 \\
V9 &= X00X1
\end{align*}
\]

Figure 1. (a) Test set for the c17 benchmark (b) corresponding expanded lists (c) decimal notation

\[
\begin{align*}
V & \text{ Including list of } V & V & \text{ Including list of } V \\
5 & [L8] & 20 [L1] \\
8 & [L2, L5] & 21 [L1, L8] \\
\end{align*}
\]

Figure 2. Including lists of the vectors of the expanded lists of Figure 1(b)

Definition 3: Consider a test set T with t vectors Tt, 1≤i≤t, containing don’t care values. The covering list of the test set T is a list, each element of which is a vector belonging to one (or more) of the expanding lists of the vectors Ti that belong to T, (in ascending order) together with the including list of each test vector.

For example, the covering list for the test set under consideration is presented in Figure 3.
Lemma 2: For $V_3$ for $V_1$ then, if the pair $(V_1, V_t)$ does not cover the test set, no pair $(V_2, V_3)$, since $V_2 \geq V_1$. Q.E.D.

For example, in Table I, the pair $(16,29)$ does not cover the test set. Therefore, none of the pairs whose first vector is $V_2$, with $V_2 \geq 16$ can cover the test set. These “hopeless” pairs are presented in the Table 1.

3. THE PROPOSED ALGORITHM

Given the definitions of the previous Section, the proposed TEX algorithm operates in 3 steps as outlined below:

Algorithm TEX $(T, t, n)$

// $T$ is a test set, containing $t$ test vectors $T_i$, $(1 \leq i \leq t)$. Each vector in the test set contains $n$-tuples of ‘1’, ‘0’ and ‘X’.

Step 1. Construct the expanded list $L_i$ for every test vector $T_i$ in $T$.

Step 2. Construct $CL$, the covering list of the test set.

Step 3. For all pairs of numbers $(V_i, V_j)$ in $CL$, calculate the lists covered by the pair $(V_i, V_j)$. Calculate the distance of all pairs covering the test set $T$. Among the pairs that cover the test set, the pair having the smallest distance is chosen as the best solution for the considered generator.

The following Observations can be used to drastically reduce the time required by the TEX algorithm. For a justification of Observation 1, see Lemma 1; for a justification of Observation 2, see Lemma 2.

Observation 1: If pair $(V_i, V_j)$ generates all lists, there is no point in examining pairs $(V_e, V_k)$ for $k \geq j$, since the distance will be greater.

For example, in Table I, the pair $(13,24)$ covers the test set; thus, there is no reason to examine pairs $(13,25), (13,26)$ etc., since the respective distance $(D)$ will be certainly greater. Taking into account Observation 1, the number of tried pairs in this example reduces by $71$ (grey shaded vectors in Table I).

Observation 2: Let $V_n$ be the greatest vector that belongs to the expanded lists of all vectors of the test set. If the pair $(V_i, V_n)$ does not generate the test set, no pair $(V_j, V_k)$ for $V_j \geq V_i, V_k \geq V_j$, will cover the test set.

For example, in Table I, since the pair $(16,29)$ (last try having 16 as the first vector) does not generate all lists, there is no point in trying pairs whose first vector is 17, 18, ... With Observation 2, the number of tried pairs is further reduced by another 78 pairs.

Therefore, thanks to observations (1) and (2) a total reduction of $71 + 78 = 149 / 325 \approx 45\%$ is achieved.

In order to perform an analysis of the proposed algorithm, we shall calculate the complexity of each step as follows.

Step 1. Requires the generation of the including list $L_i$ of each vector of the test set. If $t$ is the number of unspecified (don’t care) bits of a test vector, the complexity of this step is of the order $O(2^t)$. Therefore, the complexity of the step is of the order

$$\sum_{i=1}^{t} 2^{L_i} \quad (1)$$
The complexity of this step is of the order $O(n\log n)$ required by the insertion sort algorithm to insert the patterns into the sorted list.

Step 3. The complexity of this step is of the order $O(n^2)$, where $n$ is the number of test patterns in the covering list. Since the number of test patterns depends on the number of the unspecified bits in the original test set, the complexity is upper bounded by $O(\left(\sum_{i=1}^{k} 2^{k_i}\right)^2)$.

The upper bound lies on the fact that some (if not many) vectors are contained in the lists of more than one vector, therefore the number of patterns in this step is (much) less than the one calculated by (1).

4. SIMULATIONS AND EXPERIMENTS

In order to investigate the applicability of the proposed algorithm, we have applied it into randomly-generated test sets for various values of $n$, the number of bits and for various values of the number of $X$, the don’t care values. The utilized test sets are presented in the Appendix.

In Table 2 we present the results of the conducted experiments. Initially we generated randomly test patterns containing unspecified bits of various values of $n$ and computed the length of the sequence required by a binary counter to generate the test patterns. Then we filled the unspecified values with 0 and 1, in three ways: all $X$-values were substituted by 0 (column denoted ‘All-0’) or 1 (column denoted ‘All-1’), or the unspecified bits were filled randomly (column denoted ‘Random’). We calculated the length of the sequence required to generate the resulting test set in each case.

In Table 2 in the first column we present the value of the number of the bits ($n$). In the second column we present the length of the sequence calculated by the proposed scheme. In the next columns we present the length of the sequence required to generate the patterns resulting after substituting ‘X’ with either ‘0’ (second column), ‘1’ (third column) or randomly ‘1’ or ‘0’ (fourth column), as well as the increase in the number of required cycles. From Table 2 we can conclude that not only the proposed scheme always results in a shorter test sequence than all the other sequences (as expected), but also this reduction increases as the number of bits increases.

It should be noted that the presented results come as a straightforward application of the proposed scheme into the sequence generated by a simple binary counter. However, the proposed algorithm can be applied to a wide variety of BIST generators, e.g. counter with shifted outputs, accumulators accumulating a constant value, or even the -ubiquitous- LFSRs.

5. CONCLUSIONS

The problem of reliability of current electronic systems has been the subject of various researchers, e.g. [1]. Mobile devices call for an unprecedented need for low power during both normal operation and testing. Reducing the length of the required test patterns reduces both test time and consumed power. Test set embedding schemes [4]-[9] utilize generators that either exist or can be easily implemented on-chip and try to embed complete test sets within a subsequence (as short as possible) of the sequence generated by the utilized hardware. Test sets extracted from well-known vendor tools contain don’t-care (i.e. ‘X’) values which can be set to ‘0’ or ‘1’ in order to drive down the number of test patterns.

In this paper, a novel algorithm has been presented that can be utilized in order to embed test sets containing don’t-care ($X$) values into sequences generated by binary counters. Conducted experiments reveal that the proposed scheme results in considerably shorter sequences compared to filling the $X$ values with all-‘0’, all-‘1’ or random filling. Therefore, both test time and consumed power can be considerably reduced. The proposed scheme may prove useful for the testing of mobile devices.

6. REFERENCES


Table 1. Hopeless vectors for the example test set

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<tr>
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<th>17,18</th>
<th>17,19</th>
<th>17,20</th>
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<th>17,23</th>
<th>17,24</th>
<th>17,25</th>
<th>17,26</th>
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Table 2. Considered test sets and results

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<td>824</td>
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<td>511</td>
<td>54%</td>
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<td>73%</td>
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Appendix: Test sets utilized in our experiments

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